

## 3.3-V PHASE-LOCK LOOP CLOCK DRIVER WITH POWER DOWN MODE

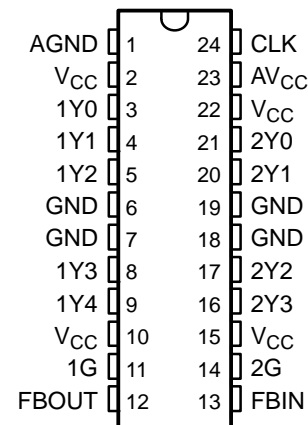
### FEATURES

- Designed to Meet and Exceed PC133 SDRAM Registered DIMM Specification Rev. 1.1
- Spread Spectrum Clock Compatible
- Operating Frequency 20 MHz to 175 MHz
- Static Phase Error Distribution at 66 MHz to 166 MHz Is  $\pm 125$  ps
- Jitter (cyc - cyc) at 60 MHz to 175 MHz Is Typ = 65 ps
- Advanced Deep Submicron Process Results in More Than 40% Lower Power Consumption Versus Current Generation PC133 Devices
- Auto Frequency Detection to Disable Device (Power-Down Mode)
- Available in Plastic 24-Pin TSSOP
- Phase-Lock Loop Clock Distribution for Synchronous DRAM Applications
- Distributes One Clock Input to One Bank of Five and One Bank of Four Outputs
- Separate Output Enable for Each Output Bank
- External Feedback (FBIN) Terminal Is Used to Synchronize the Outputs to the Clock Input
- 25- $\Omega$  On-Chip Series Damping Resistors
- No External RC Network Required
- Operates at 3.3 V

### APPLICATIONS

- DRAM Applications
- PLL Based Clock Distributors
- Non-PLL Clock Buffer

PW PACKAGE  
(TOP VIEW)



### DESCRIPTION

The CDCVF2509A is a high-performance, low-skew, low-jitter, phase-lock loop (PLL) clock driver. It uses a PLL to precisely align, in both frequency and phase, the feedback (FBOUT) output to the clock (CLK) input signal. It is specifically designed for use with synchronous DRAMs. The CDCVF2509A operates at a 3.3-V V<sub>CC</sub>. It also provides integrated series-damping resistors that make it ideal for driving point-to-point loads.

One bank of five outputs and one bank of four outputs provide nine low-skew, low-jitter copies of CLK. Output signal duty cycles are adjusted to 50%, independent of the duty cycle at CLK. Each bank of outputs is enabled or disabled separately via the control (1G and 2G) inputs. When the G inputs are high, the outputs switch in phase and frequency with CLK; when the G inputs are low, the outputs are disabled to the logic-low state. The device automatically goes into power-down mode when no input signal (< 1 MHz) is applied to CLK; the outputs go into a low state.

Unlike many products containing PLLs, the CDCVF2509A does not require external RC networks. The loop filter for the PLL is included on-chip, minimizing component count, board space, and cost.



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Because it is based on PLL circuitry, the CDCVF2509A requires a stabilization time to achieve phase lock of the feedback signal to the reference signal. This stabilization time is required following power up and application of a fixed-frequency, fixed-phase signal at CLK, and following any changes to the PLL reference or feedback signals. The PLL can be bypassed by strapping AV<sub>CC</sub> to ground to use as a simple clock buffer.

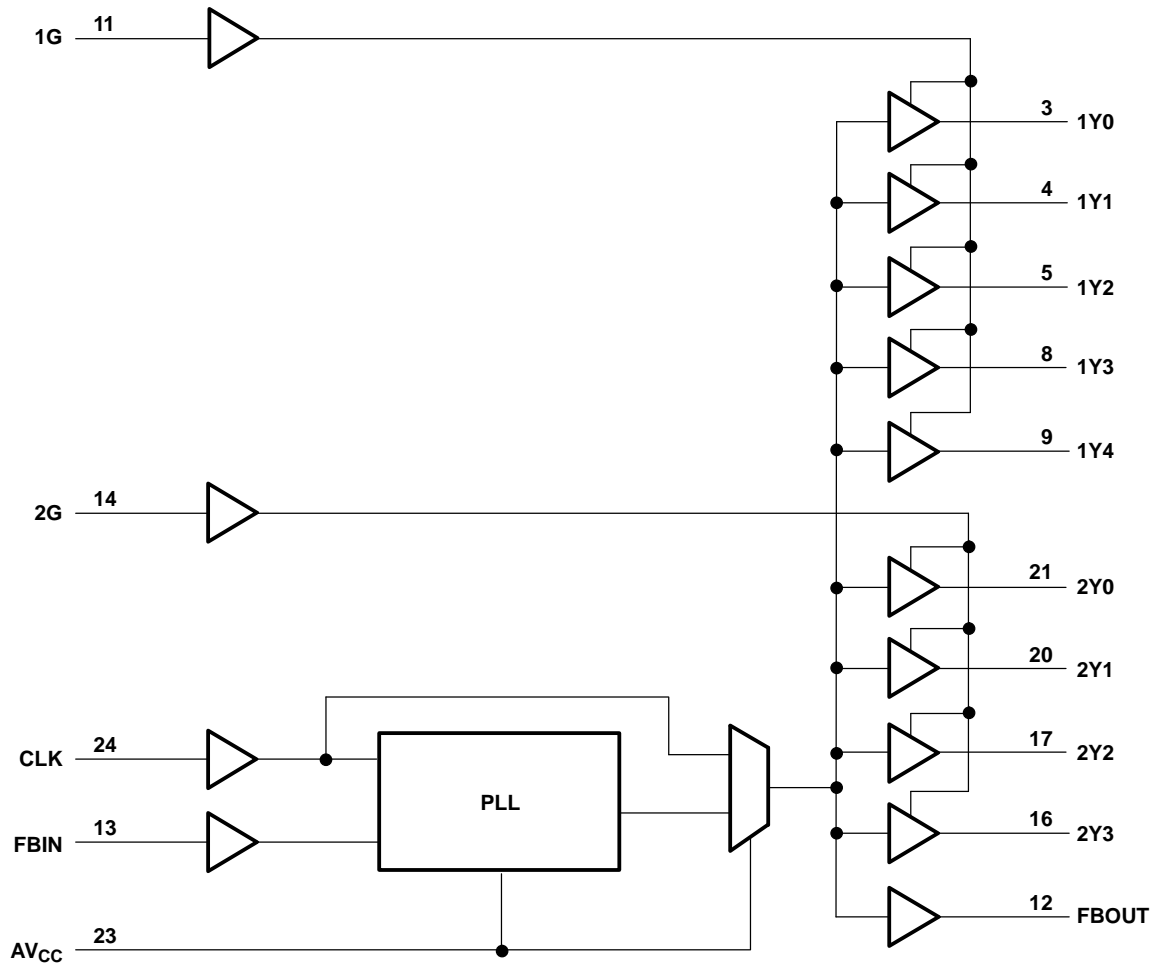
The CDCVF2509A is characterized for operation from 0°C to 85°C.

For application information, see application reports *High Speed Distribution Design Techniques for CDC509/516/2509/2510/2516* (SLMA003) and *Using CDC2509A/2510A PLL with Spread Spectrum Clocking (SSC)* (SCAA039).

**FUNCTION TABLE**

Inputs			Outputs		PLL
AVDD	1G/2G	CLK	1Y/2Y	FBOUT	
GND	H	L	L	L	Bypassed / Off
GND	H	H	H	H	Bypassed / Off
GND	L	L	L	L	Bypassed / Off
GND	L	H	L	H	Bypassed / Off
GND	L	Toggling	L	Toggling in phase to CLK	Bypassed / Off
3.3 V (nom)	L	H	L	L	On
3.3 V (nom)	L	Toggling	L	Toggling in phase to CLK	On
3.3 V (nom)	H	L	L	L	On
3.3 V (nom)	H	H	H	H	On
3.3 V (nom)	H	Toggling	Toggling in phase to CLK	Toggling in phase to CLK	On
3.3 V (nom)	X	< 1 MHz	L	L	Off

**FUNCTIONAL BLOCK DIAGRAM**



**AVAILABLE OPTIONS**

T <sub>A</sub>	PACKAGE
	0°C to 85°C
CDCVF2509APWR	
CDCVF2509APW	

### Terminal Functions

TERMINAL NAME	TERMINAL NO.	TYPE	DESCRIPTION
CLK	24	I	Clock input. CLK provides the clock signal to be distributed by the CDCVF2509A clock driver. CLK is used to provide the reference signal to the integrated PLL that generates the clock output signals. CLK must have a fixed frequency and fixed phase for the PLL to obtain phase lock. Once the circuit is powered up and a valid CLK signal is applied, a stabilization time is required for the PLL to phase lock the feedback signal to its reference signal.
FBIN	13	I	Feedback input. FBIN provides the feedback signal to the internal PLL. FBIN must be hard-wired to FBOUT to complete the PLL. The integrated PLL synchronizes CLK and FBIN so that there is nominally zero phase error between CLK and FBIN.
1G	11	I	Output bank enable. 1G is the output enable for outputs 1Y(0:4). When 1G is low, outputs 1Y(0:4) are disabled to a logic-low state. When 1G is high, all outputs 1Y(0:4) are enabled and switch at the same frequency as CLK.
2G	14	I	Output bank enable. 2G is the output enable for outputs 2Y(0:3). When 2G is low, outputs 2Y(0:3) are disabled to a logic low state. When 2G is high, all outputs 2Y(0:3) are enabled and switch at the same frequency as CLK.
FBOUT	12	O	Feedback output. FBOUT is dedicated for external feedback. It switches at the same frequency as CLK. When externally wired to FBIN, FBOUT completes the feedback loop of the PLL. FBOUT has an integrated 25- $\Omega$ series-damping resistor.
1Y (0:4)	3, 4, 5, 8, 9	O	Clock outputs. These outputs provide low-skew copies of CLK. Output bank 1Y(0:4) is enabled via the 1G input. These outputs can be disabled to a logic-low state by deasserting the 1G control input. Each output has an integrated 25- $\Omega$ series-damping resistor.
2Y (0:3)	16, 17, 21, 20	O	Clock outputs. These outputs provide low-skew copies of CLK. Output bank 2Y(0:3) is enabled via the 2G input. These outputs can be disabled to a logic-low state by deasserting the 2G control input. Each output has an integrated 25- $\Omega$ series-damping resistor.
AV <sub>CC</sub>	23	Power	Analog power supply. AV <sub>CC</sub> provides the power reference for the analog circuitry. In addition, AV <sub>CC</sub> can be used to bypass the PLL. When AV <sub>CC</sub> is strapped to ground, PLL is bypassed and CLK is buffered directly to the device outputs.
AGND	1	Ground	Analog ground. AGND provides the ground reference for the analog circuitry.
V <sub>CC</sub>	2, 10, 15, 22	Power	Power supply
GND	6, 7, 18, 19	Ground	Ground

### ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

	UNIT
AV <sub>CC</sub> Supply voltage range <sup>(2)</sup>	AV <sub>CC</sub> < V <sub>CC</sub> + 0.7 V
V <sub>CC</sub> Supply voltage range	–0.5 V to 4.3 V
V <sub>I</sub> Input voltage range <sup>(3)</sup>	–0.5 V to 4.6 V
V <sub>O</sub> Voltage range applied to any output in the high or low state <sup>(3)(4)</sup>	–0.5 V to V <sub>CC</sub> + 0.5 V
I <sub>IK</sub> Input clamp current (V <sub>I</sub> < 0)	–50 mA
I <sub>OK</sub> Output clamp current (V <sub>O</sub> < 0 or V <sub>O</sub> > V <sub>CC</sub> )	±50 mA
I <sub>O</sub> Continuous output current (V <sub>O</sub> = 0 to V <sub>CC</sub> )	±50 mA
Continuous current through each V <sub>CC</sub> or GND	±100 mA
Maximum power dissipation at T <sub>A</sub> = 55°C (in still air) <sup>(5)</sup>	0.7 W
T <sub>stg</sub> Storage temperature range	–65°C to 150°C

- (1) Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) AV<sub>CC</sub> **must not** exceed V<sub>CC</sub> + 0.7 V
- (3) The input and output negative-voltage ratings may be exceeded if the input and output clamp-current ratings are observed.
- (4) This value is limited to 4.6 V maximum.
- (5) The maximum package power dissipation is calculated using a junction temperature of 150°C and a board trace length of 750 mils. For more information, see the *Package Thermal Considerations* application note in the *ABT Advanced BiCMOS Technology Data Book* (SCBD002).

## RECOMMENDED OPERATING CONDITIONS<sup>(1)</sup>

		MIN	MAX	UNIT
$V_{CC}, AV_{CC}$	Supply voltage	3	3.6	V
$V_{IH}$	High-level input voltage	2		V
$V_{IL}$	Low-level input voltage		0.8	V
$V_I$	Input voltage	0	$V_{CC}$	V
$I_{OH}$	High-level output current		-12	mA
$I_{OL}$	Low-level output current		12	mA
$T_A$	Operating free-air temperature	0	85	°C

(1) Unused inputs must be held high or low to prevent them from floating.

## TIMING REQUIREMENTS

over recommended ranges of supply voltage and operating free-air temperature

		MIN	MAX	UNIT
$f_{clk}$	Clock frequency	20	175	MHz
	Input clock duty cycle	40%	60%	
	Stabilization time <sup>(1)</sup>		1	ms

(1) The time required for the integrated PLL circuit to obtain phase lock of its feedback signal to its reference signal. For phase lock to be obtained, a fixed-frequency, fixed-phase reference signal must be present at CLK. Until phase lock is obtained, the specifications for propagation delay, skew and jitter parameters given in the *switching characteristics* table are not applicable. This parameter does not apply for input modulation under SSC application.

## ELECTRICAL CHARACTERISTICS

over recommended operating free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS	$V_{CC}, AV_{CC}$	MIN	TYP <sup>(1)</sup>	MAX	UNIT
$V_{IK}$	Input clamp voltage	$I_I = -18$ mA	3		-1.2	V
$V_{OH}$	High-level output voltage	$I_{OH} = -100$ $\mu$ A	MIN to MAX	$V_{CC}-0.2$		V
		$I_{OH} = -12$ mA	3	2.1		
		$I_{OH} = -6$ mA	3	2.4		
$V_{OL}$	Low-level output voltage	$I_{OL} = 100$ $\mu$ A	MIN to MAX		0.2	V
		$I_{OL} = 12$ mA	3		0.8	
		$I_{OL} = 6$ mA	3		0.55	
$I_{OH}$	High-level output current	$V_O = 1$ V	3		-28	mA
		$V_O = 1.65$ V	3.3		-36	
		$V_O = 3.135$ V	3.6		-8	
$I_{OL}$	Low-level output current	$V_O = 1.95$ V	3	30		mA
		$V_O = 1.65$ V	3.3	40		
		$V_O = 0.4$ V	3.6		10	
$I_I$	Input current	$V_I = V_{CC}$ or GND	3.6		$\pm 5$	$\mu$ A
$I_{CC}$ <sup>(2)</sup>	Supply current (static, output not switching)	$V_I = V_{CC}$ or GND, $I_O = 0$ , Outputs: low or high	3.6	0	40	$\mu$ A
$\Delta I_{CC}$	Change in supply current	One input at $V_{CC} - 0.6$ V, Other inputs at $V_{CC}$ or GND	3.3	3.6	500	$\mu$ A
$C_i$	Input capacitance	$V_I = V_{CC}$ or GND	3.3		2.5	pF
$C_o$	Output capacitance	$V_O = V_{CC}$ or GND	3.3		2.8	pF

(1) For conditions shown as MIN or MAX, use the appropriate value specified under the *recommended operating conditions* section.

(2) For dynamic  $I_{CC}$  vs Frequency, see [Figure 9](#) and [Figure 10](#).

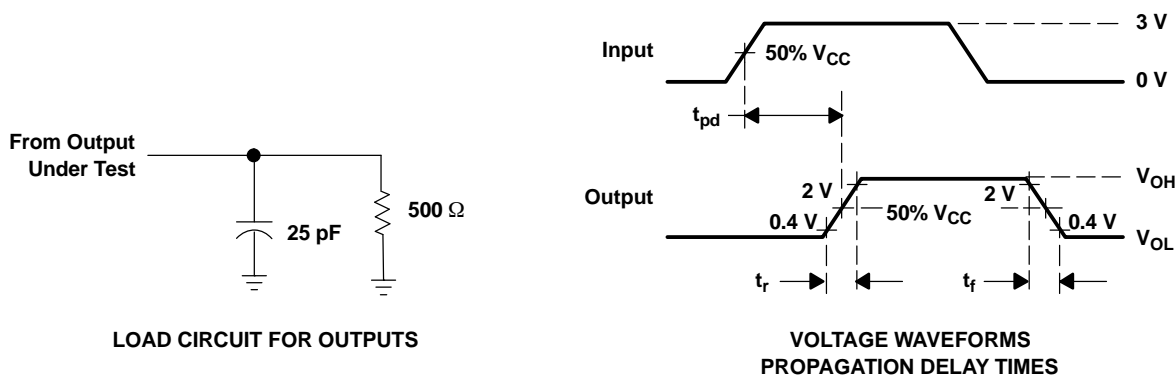
### SWITCHING CHARACTERISTICS

over recommended ranges of supply voltage and operating free-air temperature,  $C_L = 25\text{ pF}$  (see Figure 1 and Figure 2)<sup>(1)(2)</sup>

PARAMETER	FROM (INPUT)	TO (OUTPUT)	$V_{CC}, AV_{CC} = 3.3\text{ V} \pm 0.3\text{ V}$			UNIT
			MIN	TYP	MAX	
$t_{\phi}$ Phase error time- static (normalized) (see Figure 4 through Figure 7)	CLK $\uparrow$ = 25 MHz to 65 MHz CLK $\uparrow$ = 66 MHz to 166 MHz	FBIN $\uparrow$	-150 -125		150 125	ps
$t_{sk(o)}$ Output skew time <sup>(3)</sup>	Any Y	Any Y			100	ps
Phase error time-jitter <sup>(4)</sup>	CLK = 66 MHz to 100 MHz	Any Y or FBOUT	-50		50	ps
$Jitter_{(cycle-cycle)}$ (see Figure 8)	CLK = 25 MHz to 40 MHz	Any Y or FBOUT			500	ps
	CLK = 41 MHz to 59 MHz				200	
	CLK = 60 MHz to 175 MHz			65	125	
$t_{d(o)}$ Dynamic phase offset <sup>(5)</sup>	CLK $\uparrow$ = 25 MHz to 65 MHz	FBIN $\uparrow$			1.5	ns
	CLK $\uparrow$ = 66 MHz to 166 MHz				0.4	
Duty cycle	$f_{(CLK)} > 60\text{ MHz}$	Any Y or FBOUT	45%		55%	
$t_r$ Rise time	$V_O = 0.4\text{ V to }2\text{ V}$	Any Y or FBOUT	0.3		1.1	ns/V
$t_f$ Fall time	$V_O = 2\text{ V to }0.4\text{ V}$	Any Y or FBOUT	0.3		1.1	ns/V
$t_{PLH}$ Low-to-high propagation delay time, bypass mode	CLK	Any Y or FBOUT	1.8		3.9	ns
$t_{PHL}$ High-to-low propagation delay time, bypass mode	CLK	Any Y or FBOUT	1.8		3.9	ns

- (1) The specifications for parameters in this table are applicable only after any appropriate stabilization time has elapsed.
- (2) These parameters are not production tested.
- (3) The  $t_{sk(o)}$  specification is only valid for equal loading of all outputs.
- (4) Calculated per PC DRAM SPEC ( $t_{phase\ error}, static-jitter_{(cycle-to-cycle)}$ ).
- (5) The parameter is assured by design but cannot be 100% production tested.

### PARAMETER MEASUREMENT INFORMATION



- NOTES:
- A.  $C_L$  includes probe and jig capacitance.
  - B. All input pulses are supplied by generators having the following characteristics: PRR  $\leq 133\text{ MHz}$ ,  $Z_O = 50\ \Omega$ ,  $t_r \leq 1.2\text{ ns}$ ,  $t_f \leq 1.2\text{ ns}$ .
  - C. The outputs are measured one at a time with one transition per measurement.

Figure 1. Load Circuit and Voltage Waveforms

PARAMETER MEASUREMENT INFORMATION (continued)

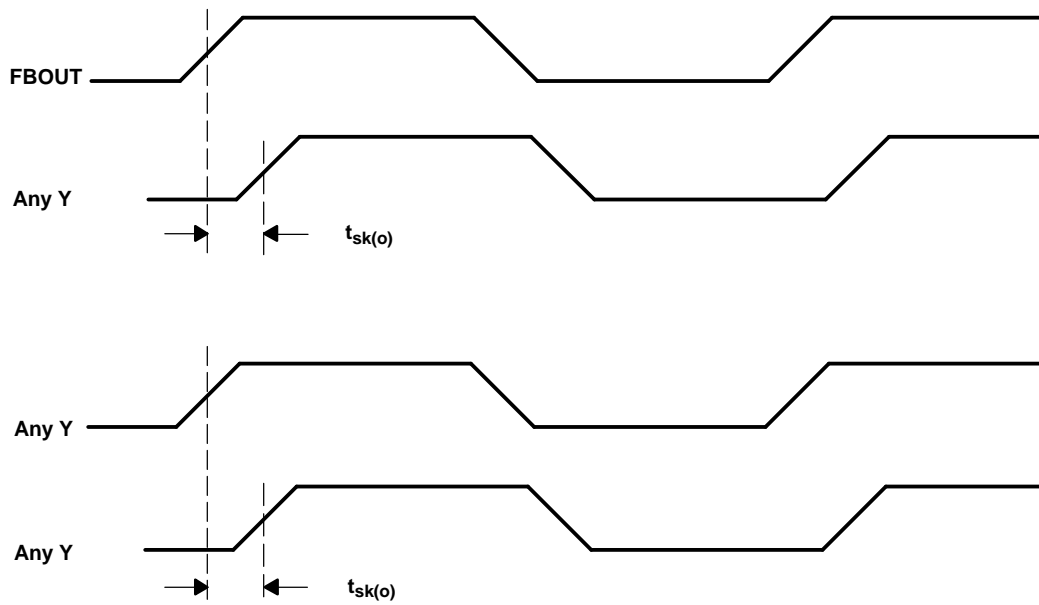
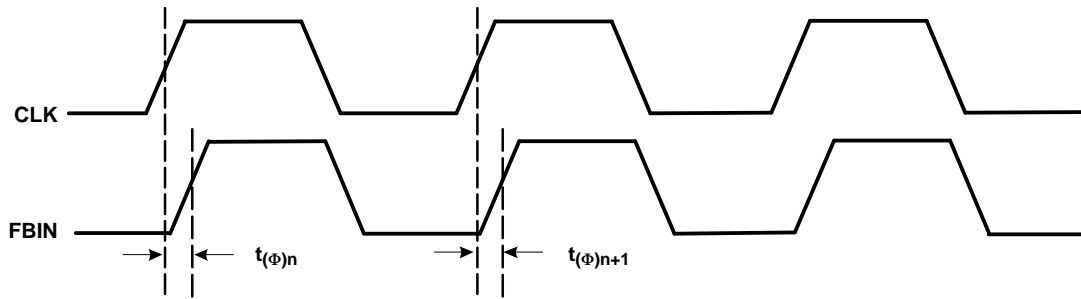


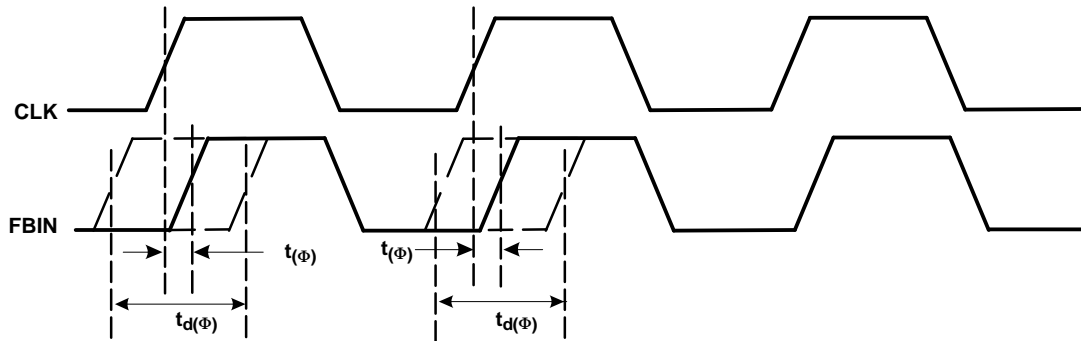
Figure 2. Skew Calculations

PARAMETER MEASUREMENT INFORMATION (continued)



$$t_{(\phi)} = \frac{\sum_{n=1}^{n=N} t_{(\phi)n}}{N} \quad (N \text{ is a large number of samples})$$

a) Static Phase Offset



b) Dynamic Phase Offset

Figure 3. Static and Dynamic Phase Offset



**TYPICAL CHARACTERISTICS**

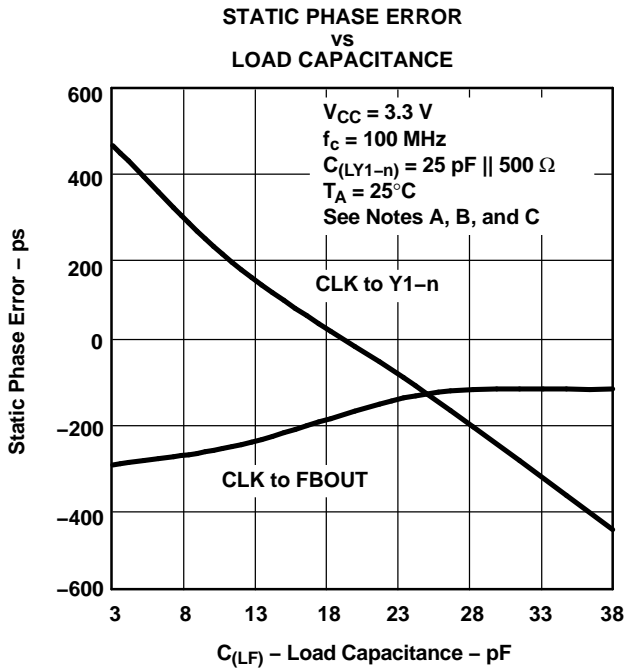


Figure 4.

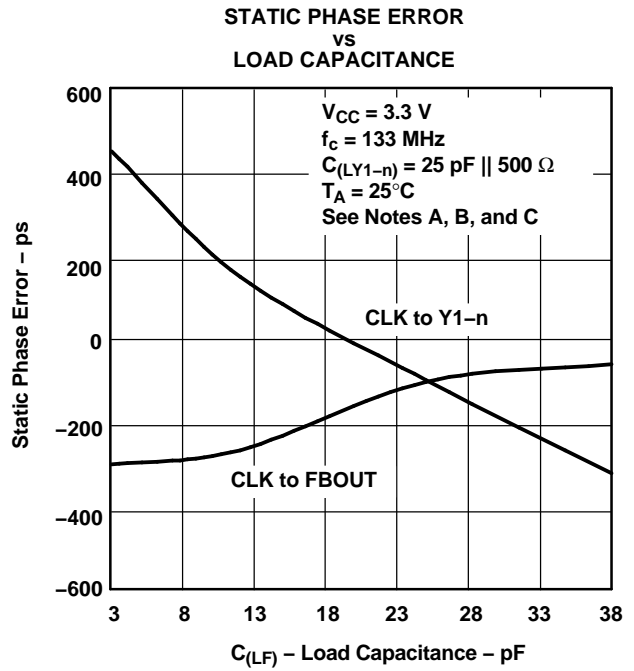


Figure 5.

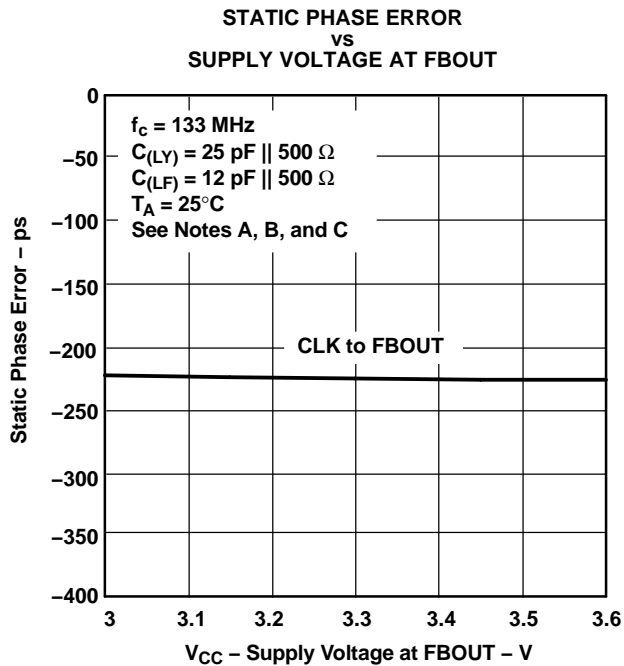


Figure 6.

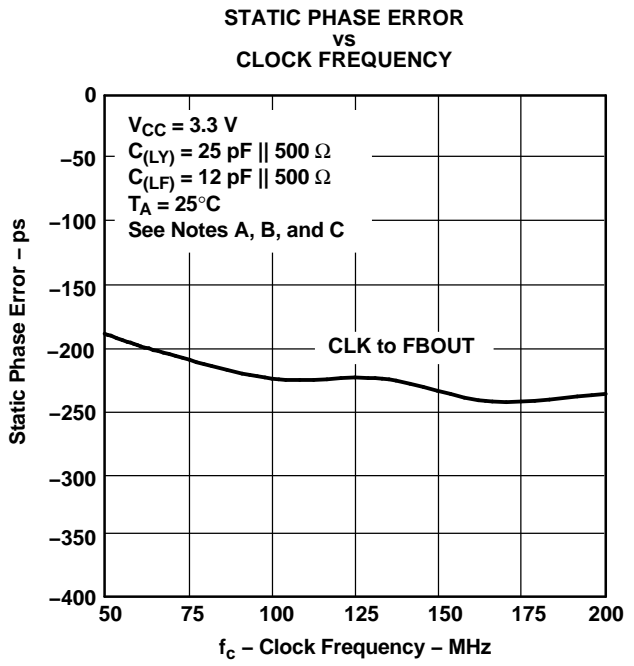


Figure 7.

- A. Trace length FBOUT to FBIN = 5 mm,  $Z_0 = 50\ \Omega$
- B.  $C_{(LY)}$  = Lumped capacitive load  $Y_{1-n}$
- C.  $C_{(LFx)}$  = Lumped feedback capacitance at FBOUT = FBIN

TYPICAL CHARACTERISTICS (continued)

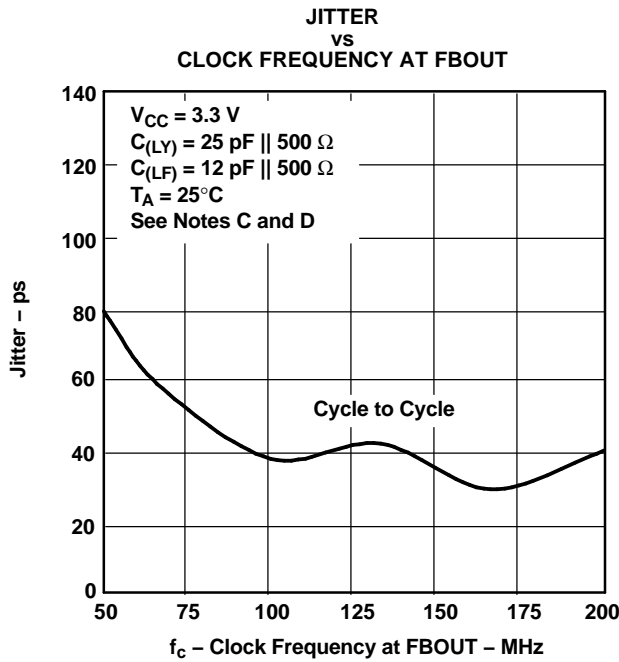


Figure 8.

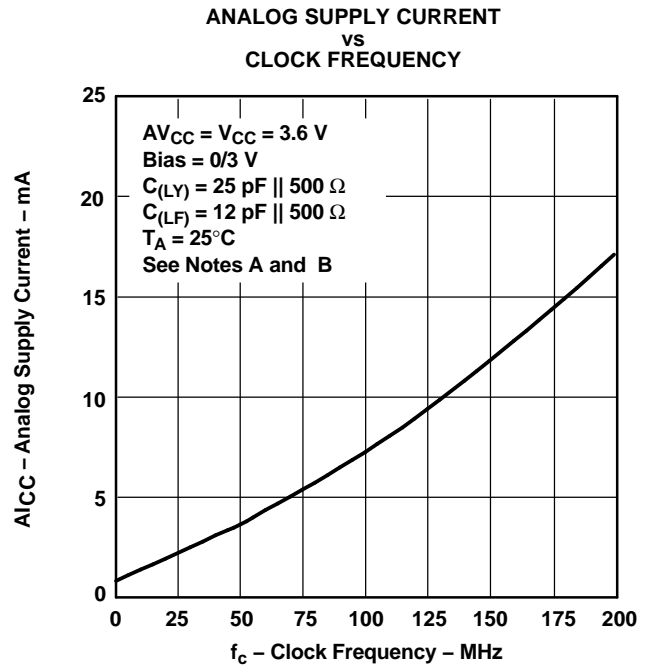


Figure 9.

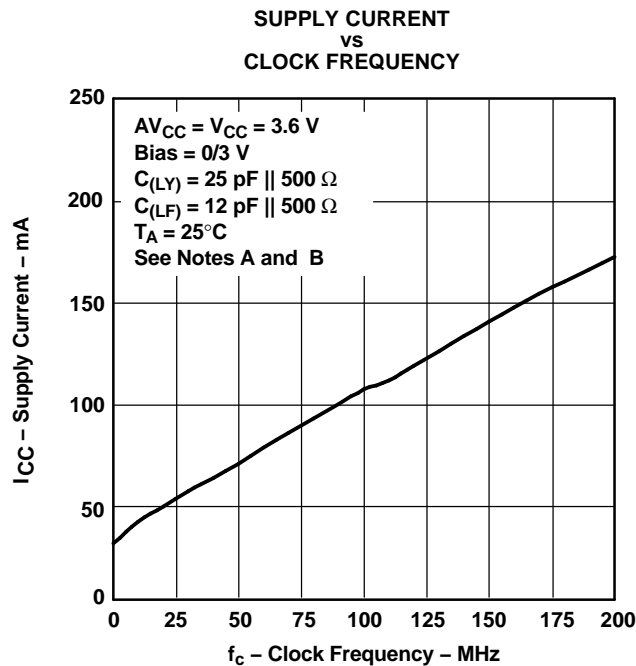


Figure 10.

- A. Trace length FBOUT to FBIN = 5 mm,  $Z_0 = 50 \Omega$
- B.  $C_{(LY)}$  = Lumped capacitive load  $Y_{1-n}$
- C.  $C_{(LFx)}$  = Lumped feedback capacitance at FBOUT = FBIN
- D.  $C_{(LFx)}$  = Lumped feedback capacitance at FBOUT = FBIN.

**Revision History**

**Table 1. Revision History**

<b>Date</b>	<b>Rev</b>	<b>Page</b>	<b>Section</b>	<b>Description</b>
04/11/05	B	6	Switching Characteristics	Added static phase error - 25 MHz to 65 MHz
				Added jitter - 25 MHz to 65 MHz
				Added Dynamic Phase Offset specification
		7	Figure 2	Revised into two figures
8	Figure 3	Added Figure 3 for a diagram of dynamic phase offset		

**TAPE AND REEL INFORMATION**



**QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE**



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
CDCVF2509APWR	TSSOP	PW	24	2000	330.0	16.4	6.95	8.3	1.6	8.0	16.0	Q1

**TAPE AND REEL BOX DIMENSIONS**



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
CDCVF2509APWR	TSSOP	PW	24	2000	346.0	346.0	33.0

PW (R-PDSO-G\*\*)

PLASTIC SMALL-OUTLINE PACKAGE

14 PINS SHOWN



4040064/F 01/97

- NOTES: A. All linear dimensions are in millimeters.  
 B. This drawing is subject to change without notice.  
 C. Body dimensions do not include mold flash or protrusion not to exceed 0,15.  
 D. Falls within JEDEC MO-153

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Microcontrollers	<a href="http://microcontroller.ti.com">microcontroller.ti.com</a>
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### Applications

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Automotive	<a href="http://www.ti.com/automotive">www.ti.com/automotive</a>
Broadband	<a href="http://www.ti.com/broadband">www.ti.com/broadband</a>
Digital Control	<a href="http://www.ti.com/digitalcontrol">www.ti.com/digitalcontrol</a>
Medical	<a href="http://www.ti.com/medical">www.ti.com/medical</a>
Military	<a href="http://www.ti.com/military">www.ti.com/military</a>
Optical Networking	<a href="http://www.ti.com/opticalnetwork">www.ti.com/opticalnetwork</a>
Security	<a href="http://www.ti.com/security">www.ti.com/security</a>
Telephony	<a href="http://www.ti.com/telephony">www.ti.com/telephony</a>
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